

200mA Low-Dropout Linear Regulator with Pin-Selectable Dual-Voltage Level Output

FEATURES

- Very Low Dropout: 230mV Typical at 200mA
- 3% Accuracy Over Load/Line/Temperature
- Low I_Q: 50µA in Active Mode
- Available in Fixed-Output Voltages From 0.9V to 3.6V Using Innovative Factory EEPROM Programming
- VSET Pin Toggles Output Voltage Between Two Preset Levels
 - Preset Output Voltage Levels Can Be EEPROM-Programmed To Any Combination
- High PSRR: 65dB at 1kHz
- Stable with a 1.0µF Ceramic Capacitor
- Thermal Shutdown and Over-Current Protection
- Available in Wafer-Level Chip Scale and 2mm x 2mm SON Packages

APPLICATIONS

- Power Rails with Programming Mode
- Dual Voltage Levels for Power-Saving Mode
- Leakage Reduction for 90nm and 65nm Processors
- Wireless Handsets, Smart Phones, PDAs
- MP3 Players and Other Handheld Products

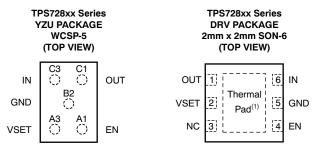
DESCRIPTION

The TPS728xx series of low-dropout linear regulators (LDOs), with a selectable dual-voltage level output, is designed specially for applications that require two levels of output voltage regulation. Programming fuses and memory cards, reducing leakage effects, and conserving power in nanometric processes are some application examples.

The VSET pin is used to select one of two output voltage levels preset through innovative factory EEPROM programming. A precision bandgap and error amplifier provides an overall 3% accuracy over load, line, and temperature extremes.

Ultra-small wafer chip scale (WCSP) and 2mm x 2mm SON packages make the TPS728xx series ideal for handheld applications.

This family of devices is fully specified over a temperature range of $T_1 = -40^{\circ}\text{C}$ to +125°C.



 It is recommended that the SON package thermal pad be connected to ground.

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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

ORDERING INFORMATION(1)

PRODUCT	V _{OUT} ⁽²⁾
	VVV is the nominal output voltage for V_{OUT1} and corresponds to V_{SET} = Low. XXX is the nominal output voltage for V_{OUT2} and corresponds to V_{SET} = High. YYY is package designator. Z is Tape and reel quantity (R = 3000, T = 250).

- (1) For the most current package and ordering information see the Package Option Addendum at the end of this document, or see the TI website at www.ti.com.
- (2) Output voltages from 0.9V to 3.6V in 50mV increments are available through the use of innovative factory EEPROM programming; minimum order quantities may apply. Contact factory for details and availability.

ABSOLUTE MAXIMUM RATINGS(1)

At $T_{\perp} = -40^{\circ}$ C to +125°C (unless otherwise noted). All voltages are with respect to GND.

PARAMETER		TPS728xx Series	UNIT	
Input voltage ran	ge, V _{IN}	-0.3 to +7.0	V	
Enable and VSE	T voltage range, V _{EN} and V _{SET}	-0.3 to V _{IN} + 0.3 ⁽²⁾	V	
Output voltage ra	ange, V _{OUT}	-0.3 to +7.0	V	
Maximum output	current, I _{OUT}	Internally limited		
Output short-circ	uit duration	Indefinite		
Total continuous	power dissipation, P _{DISS}	See Dissipation Ratings Table		
FOD weller	Human body model (HBM)	2	kV	
ESD rating	Charged device model (CDM)	500	V	
Operating junction	on temperature range, T _J	-55 to +150	°C	
Storage tempera	ture range, T _{STG}	-55 to +150	°C	

⁽¹⁾ Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.

DISSIPATION RATINGS

BOARD	PACKAGE	R _{eJC}	R _{0JA}	DERATING FACTOR ABOVE T _A = +25°C	T _A < +25°C	T _A = +70°C	T _A = +85°C
High-K ⁽¹⁾	DRV	20°C/W	65°C/W	15.4mW/°C	1540mW	845mW	615mW
High-K ⁽¹⁾	YZU	85°C/W	268°C/W	3.7mW/°C	370mW	205mW	150mW

(1) The JEDEC high-K (2s2p) board used to derive this data was a 3- × 3-inch, multilayer board with 1-ounce internal power and ground planes and 2-ounce copper traces on top and bottom of the board.

⁽²⁾ V_{EN} and V_{SET} absolute maximum rating is V_{IN} + 0.3V or +7.0V, whichever is less.



ELECTRICAL CHARACTERISTICS

Over operating temperature range (T_J = -40° C to +125°C), V_{IN} = V_{OUT(TYP)} + 0.5V or 2.7V, whichever is greater; I_{OUT} = 0.5mA, V_{SET} = V_{EN} = V_{IN}, C_{OUT} = 1.0 μ F, unless otherwise noted. Typical values are at T_J = +25°C.

	PARAMETER			TEST CONDITIONS			MAX	UNIT
V _{IN}	Input voltage range				2.7		6.5	V
		Nominal	$T_J = +25^{\circ}C$, V_{SET}	= high/low	-2.5		+2.5	mV
V _{OUT} ⁽¹⁾	DC output accuracy	Over V _{IN} , I _{OUT} , temperature	$V_{OUT} + 0.5V \le V_{II}$ $0mA \le I_{OUT} \le 200$	$_{N} \le 6.5 \text{V},$ $_{M} \text{MMA}, \text{V}_{\text{SET}} = \text{high/low}$	-3.0		+3.0	%
ΔV_{OUT}	Load transient		100μA to 200mA 200mA to 100μA	in 1 μ s, in 1 μ s, C _{OUT} = 1 μ F		±60.0		mV
Vo	Output voltage range				0.9		3.6	V
$\Delta V_{O}/\Delta V_{IN}$	Line regulation		$V_{OUT(NOM)} + 0.5V$ $I_{OUT} = 5mA$	\leq V _{IN} \leq 6.5V,		130		μV/V
$\Delta V_O/\Delta I_{OUT}$	Load regulation		0mA ≤ I _{OUT} ≤ 200	lmA		75		μV/mA
V_{DO}	Dropout voltage ⁽²⁾		$V_{IN} = V_{OUT(NOM)} -$	- 0.1V, I _{OUT} = 200mA		230	400	mV
I_{CL}	Output current limit		$V_{OUT} = 0.9 \times V_{OU}$	T(NOM)	240	340	575	mA
1	Ground pin current		$I_{OUT} = 0mA$			50	80	μΑ
I _{GND}	Ground pin current		$I_{OUT} = 200 \text{mA}$			120		μΑ
I _{SHDN}	Shutdown current (I _{GND})		$V_{EN} \le 0.4V, 2.7V$ $T_{J} = -40^{\circ}C \text{ to } +85$			0.10	1.0	μΑ
				f = 100Hz		65		dB
DCDD	Dawar augustu rajaatis	on ratio	$V_{IN} = 3.8V,$ $V_{OUT} = 2.8V,$ $I_{OUT} = 200mA$	f = 1kHz		65		dB
PSRR	Power-supply rejection	on ratio		f = 10kHz	55			dB
				f = 100kHz		40		dB
V _N	Output noise voltage		BW = 100Hz to 100kHz, $V_{IN} = 3.3V$, $V_{OUT} = 2.8V$, $I_{OUT} = 10mA$			$75 \times V_{OUT}$		μV_{RMS}
t _{TR}	Transition time (low-t $V_{OUT} = V_{OUT_LOW}$ to $V_{OUT} = 97\% \times V_{OUT}$	V _{OUT_HIGH}	V _{OUT_LOW} = 1.8V, V _{OUT_HIGH} = 3.15V, I _{OUT} = 10mA			60		μs
t _{STR}	Startup time ⁽³⁾		$C_{OUT} = 1.0 \mu F$			160		μs
t _{SHUT}	Shutdown time ⁽⁴⁾		$R_L = \infty$, $C_{OUT} = 1$.0μF, V _{OUT} = 2.8V		180 ⁽⁵⁾		μs
V_{HI}	VSET high (output Voor enable pin high (e	_{OUT2} selected), nabled)			1.2		V_{IN}	V
V_{LO}	VSET low (output V _O or enable pin low (dis				0		0.4	V
I _{EN} , I _{VSET}	Enable and select pin currents		EN = VSET = 6.5	V		0.04	1.0	μΑ
	Undervoltage lockout		V _{IN} rising, V _{SET} =	high/low	2.38	2.51	2.65	V
UVLO	Hysteresis		V _{IN} falling, V _{SET} =	high/low		230		mV
_	The area of about do		Shutdown, tempe	erature increasing		+160		°C
ISD	T _{SD} Thermal shutdown temperature		Reset, temperatu	re decreasing		+140		°C
TJ	Operating junction te	mperature			-40		+125	°C

⁽¹⁾ The output voltage for V_{SET} = low/high is programmed at the factory.
(2) V_{DO} is not measured for devices with V_{OUT(NOM)} < 2.8V because minimum V_{IN} = 2.7V.
(3) Time from V_{EN} = 1.2V to V_{OUT} = 97% (V_{OUT(NOM)}).
(4) Time from V_{EN} = 0.4V to V_{OUT} = 5% (V_{OUT(NOM)}).
(5) See *Shutdown* in the *Application Information* section for more details.



DEVICE INFORMATION

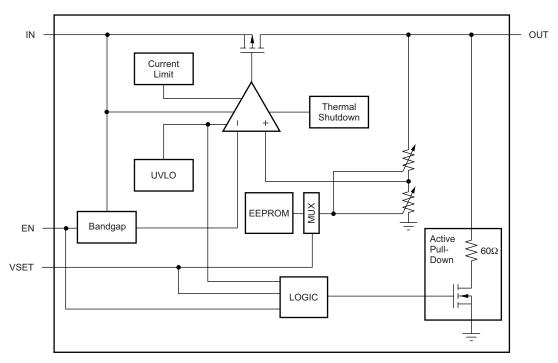
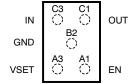


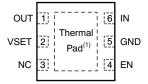
Figure 1. Functional Block Diagram



YZU PACKAGE WCSP-5 (TOP VIEW)



DRV PACKAGE SON-8 (TOP VIEW)



(1) It is recommended that the SON package thermal pad be connected to ground.

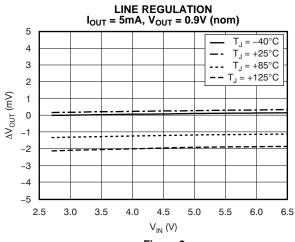
PIN DESCRIPTIONS

TPS7	28xx Seri	es	
NAME	DRV	YZU	DESCRIPTION
OUT	1	C1	Regulated output voltage pin. A small 1µF ceramic capacitor is needed from this pin to ground to assure stability. See <i>Input and Output Capacitor Requirements</i> in the <i>Application Information</i> section for more details.
VSET	2	А3	Select pin. Driving VSET below 0.4V selects preset output voltage V_{OUT1} . Driving VSET over 1.2V selects preset output voltage V_{OUT2} .
NC	3	_	No connection.
EN	4	A1	Enable pin. Driving EN over 1.2V turns on the regulator. Driving EN below 0.4V puts the regulator into shutdown mode, thus reducing the operating current to 100nA, nominal.
GND	5	B2	Ground pin (connect DRV thermal pad to ground)
IN	6	СЗ	Input pin. A small capacitor is needed from this pin to ground to assure stability. See <i>Input and Output Capacitor Requirements</i> in the <i>Application Information</i> section for more details.

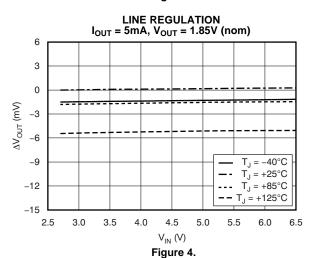


TYPICAL CHARACTERISTICS

Over operating temperature range ($T_J = -40^{\circ}C$ to $+125^{\circ}C$), $V_{IN} = V_{OUT(TYP)} + 0.5V$ or 2.7V, whichever is greater; $I_{OUT} = 0.5$ mA, $V_{EN} = V_{SET} = V_{IN}$, $C_{OUT} = 1.0 \mu F$, unless otherwise noted. Typical values are at $T_J = +25$ °C.

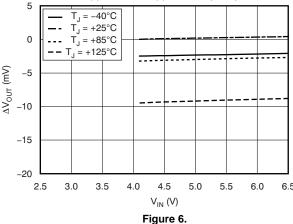






I_{OUT} = 5mA, V_{OUT} = 3.6V (nom)

LINE REGULATION



LINE REGULATION $I_{OUT} = 200$ mA, $V_{OUT} = 0.9$ V (nom)

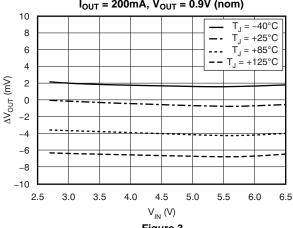
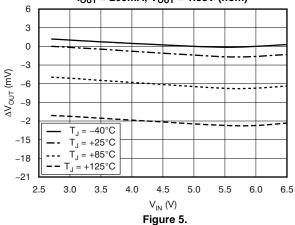


Figure 3.

LINE REGULATION I_{OUT} = 200mA, V_{OUT} = 1.85V (nom)



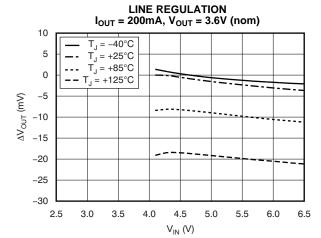
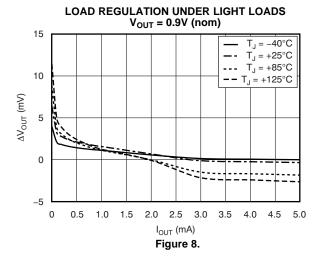
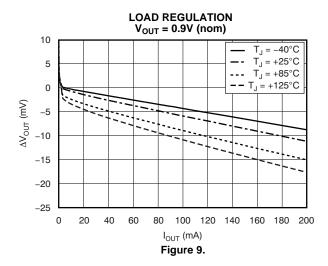


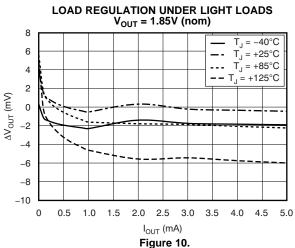
Figure 7.

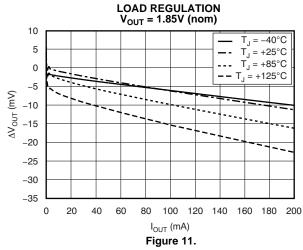


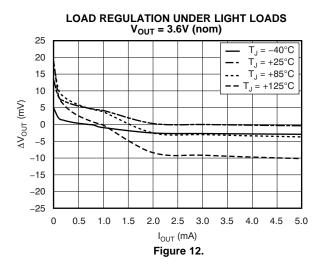
Over operating temperature range ($T_J = -40^{\circ}C$ to +125°C), $V_{IN} = V_{OUT(TYP)} + 0.5V$ or 2.7V, whichever is greater; $I_{OUT} = 0.5 \text{mA}$, $V_{EN} = V_{SET} = V_{IN}$, $C_{OUT} = 1.0 \mu F$, unless otherwise noted. Typical values are at $T_J = +25^{\circ}C$.

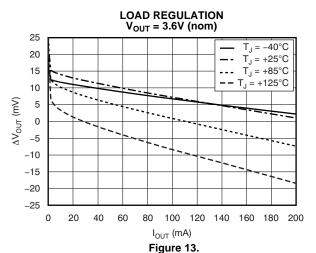














Over operating temperature range ($T_J = -40^{\circ}C$ to +125°C), $V_{IN} = V_{OUT(TYP)} + 0.5V$ or 2.7V, whichever is greater; $I_{OUT} = 0.5 \text{mA}$, $V_{EN} = V_{SET} = V_{IN}$, $C_{OUT} = 1.0 \mu F$, unless otherwise noted. Typical values are at $T_J = +25^{\circ}C$.

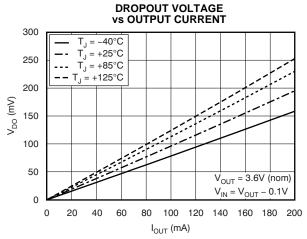


Figure 14.

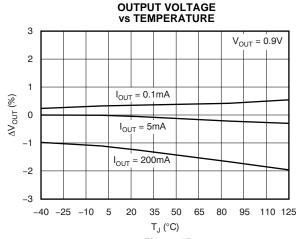


Figure 15.

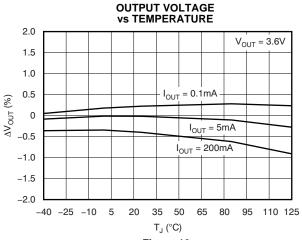


Figure 16.

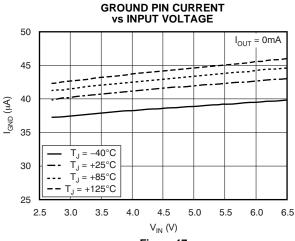
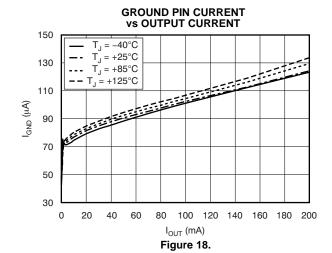


Figure 17.

GROUND PIN CURRENT

vs TEMPERATURE



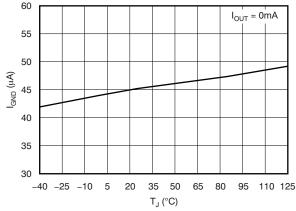
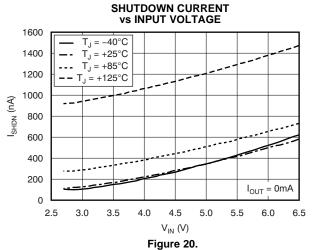


Figure 19.



Over operating temperature range (T_J = -40° C to +125°C), V_{IN} = V_{OUT(TYP)} + 0.5V or 2.7V, whichever is greater; I_{OUT} = 0.5mA, V_{EN} = V_{SET} = V_{IN}, C_{OUT} = 1.0 μ F, unless otherwise noted. Typical values are at T_J = +25°C.



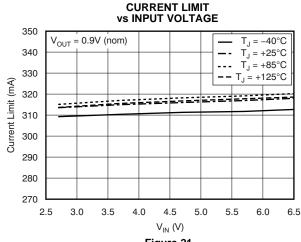


Figure 21.

CURRENT LIMIT vs INPUT VOLTAGE 350

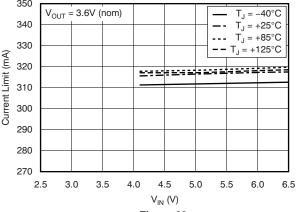


Figure 22.

TPS728185315 POWER-SUPPLY RIPPLE REJECTION vs FREQUENCY ($V_{IN} - V_{OUT} = 0.85V$) 90 $V_{IN} = 2.7V$

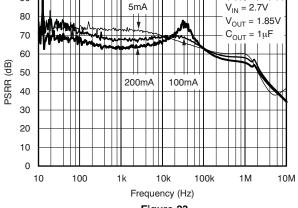


Figure 23.

TPS728185315 POWER-SUPPLY RIPPLE REJECTION vs FREQUENCY (V_{IN} – V_{OUT} = 1.0V)

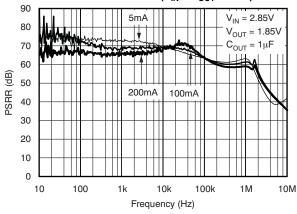


Figure 24.

TPS728185315 POWER-SUPPLY RIPPLE REJECTION vs FREQUENCY (V_{IN} - V_{OUT} = 0.5V)

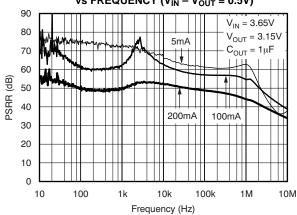


Figure 25.



Over operating temperature range ($T_J = -40^{\circ}C$ to +125°C), $V_{IN} = V_{OUT(TYP)} + 0.5V$ or 2.7V, whichever is greater; $I_{OUT} = 0.5 \text{mA}$, $V_{EN} = V_{SET} = V_{IN}$, $C_{OUT} = 1.0 \mu F$, unless otherwise noted. Typical values are at $T_J = +25^{\circ}C$.

TPS728185315 POWER-SUPPLY RIPPLE REJECTION vs FREQUENCY $(V_{IN} - V_{OUT} = 1.0V)$ 90 $V_{IN} = 4.15V$ 80 $V_{OUT} = 3.15V$ $C_{OUT} = 1\mu F$ 70 60 50 200mA 100mA 40 30 20 10 0 10 10M Frequency (Hz)

Figure 26.

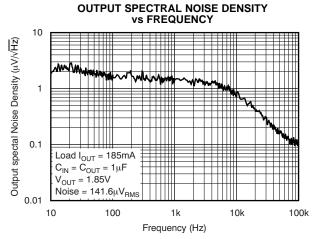


Figure 27.

OUTPUT SPECTRAL NOISE DENSITY vs FREQUENCY

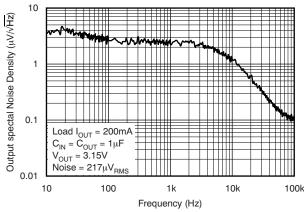


Figure 28.

LINE TRANSIENT RESPONSE

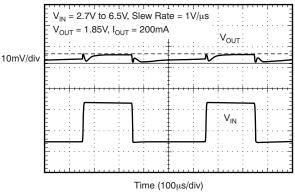


Figure 29.

LINE TRANSIENT RESPONSE

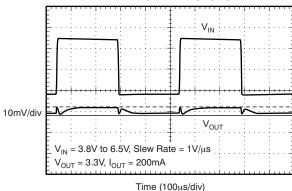
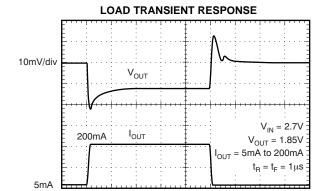


Figure 30.



Time (10µs/div) **Figure 31.**



Over operating temperature range (T_J = -40° C to +125°C), V_{IN} = V_{OUT(TYP)} + 0.5V or 2.7V, whichever is greater; I_{OUT} = 0.5mA, V_{EN} = V_{SET} = V_{IN}, C_{OUT} = 1.0 μ F, unless otherwise noted. Typical values are at T_J = +25°C.

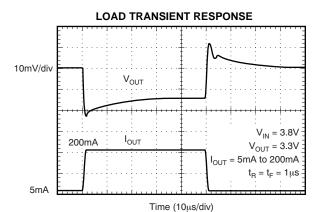


Figure 32.

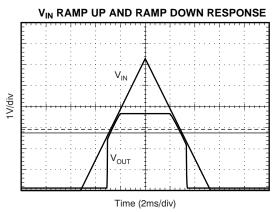


Figure 34.

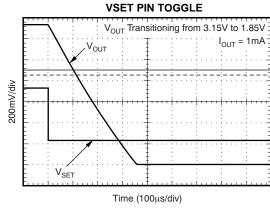


Figure 36.

ENABLE TRANSIENT RESPONSE

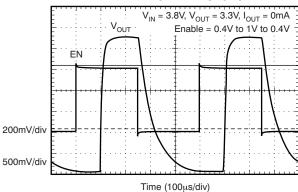
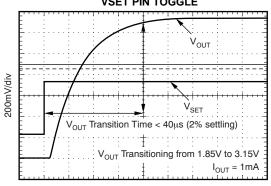


Figure 33.

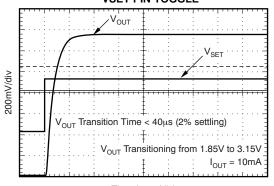
VSET PIN TOGGLE



Time (10µs/div)

Figure 35.

VSET PIN TOGGLE



Time (40µs/div) Figure 37.



Over operating temperature range (T $_J$ = -40°C to +125°C), V_{IN} = $V_{OUT(TYP)}$ + 0.5V or 2.7V, whichever is greater; I_{OUT} = 0.5mA, V_{EN} = V_{SET} = V_{IN} , C_{OUT} = 1.0 μ F, unless otherwise noted. Typical values are at T_J = +25°C.

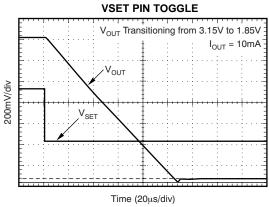


Figure 38.

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APPLICATION INFORMATION

The TPS728xx series belongs to a family of new generation LDO regulators that use innovative circuitry to achieve ultra-wide bandwidth and high loop gain, resulting in extremely high PSRR (up to 1MHz) at very low headroom ($V_{\rm IN}-V_{\rm OUT}$). These features, combined with low noise, low ground pin current, and ultra-small packaging, make this device ideal for portable applications. This family of regulators offers sub-bandgap output voltages, current limit and thermal protection, and is fully specified from -40°C to $+125^{\circ}\text{C}$.

Figure 39 shows the basic circuit connections.

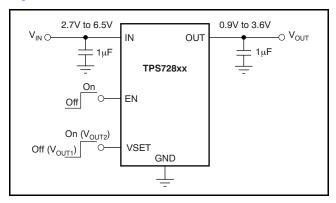


Figure 39. Typical Application Circuit

APPLICATION EXAMPLES

EEPROM-based applications require programming voltage to be higher than the operating voltage. The TPS728xx suits such applications where the maximum programming voltage of the EEPROM is higher than the operating voltage. The VSET logic pin allows the application to transition between the higher EEPROM programming voltage and the lower operating voltage. For example, the TPS728xx typically takes less than 40µs to transition from a lower voltage of 1.85V to a higher voltage of 3.15V under an output load of 1mA to 10mA, as shown in Figure 35 and Figure 37, respectively. The special circuitry in the TPS728xx helps transition from the higher voltage to the lower voltage under no load. The load on the output at the end of the programming cycle is typically under 10mA. Output voltage overshoots and undershoots are minimal under this load condition. The TPS728xx typically takes less than 1ms of transition time going from 3.15V to 1.85V, as shown in Figure 36 and Figure 38, respectively. Both output states of the TPS728xx are programmable between 0.9V to 3.6V.

Another area where the TPS728xx can be used effectively is in dynamic voltage scaling (DVS) applications. In DVS applications, it is required to dynamically switch between a high operational voltage to a low standby voltage in order to balance performance of processors and achieve power savings. Modern multimillion gate microprocessors fabricated with the latest sub-micron processes save on power by transitioning to a lower voltage to reduce leakage currents without losing content. This architecture enables the microprocessor to transition quickly into an operational state (wake up) without requiring reloading of the states from external memory, or a reboot.

INPUT AND OUTPUT CAPACITOR REQUIREMENTS

Although an input capacitor is not required for stability, it is good analog design practice to connect a 0.1µF to 1.0µF low equivalent series resistance (ESR) capacitor across the input supply near the regulator. This capacitor counteracts reactive input sources and improves transient response, noise rejection, and ripple rejection. A higher-value capacitor may be necessary if large, fast rise-time load transients are anticipated, or if the device is not located near the power source. If source impedance is not sufficiently low, a 0.1µF input capacitor may be necessary to ensure stability.

The TPS728xx is designed to be stable with standard ceramic capacitors with values of $1.0\mu\text{F}$ or larger at the output. X5R- and X7R-type capacitors are best because they have minimal variation in value and ESR over temperature. Maximum ESR should be less than 1.0Ω .

BOARD LAYOUT RECOMMENDATIONS TO IMPROVE PSRR AND NOISE PERFORMANCE

To improve ac performance such as PSRR, output noise, and transient response, it is recommended that the board be designed with separate ground planes for V_{IN} and V_{OUT} , with each ground plane connected only at the GND pin of the device. In addition, the ground connection for the output capacitor should connect directly to the GND pin of the device. High ESR capacitors may degrade PSRR.

TEXAS INSTRUMENTS

INTERNAL CURRENT LIMIT

The TPS728xx internal current limits help protect the regulator during fault conditions. During current limit, the output sources a fixed amount of current that is largely independent of output voltage. For reliable operation, the device should not be operated in a current limit state for extended periods of time.

The PMOS pass element in the TPS728xx has a built-in body diode that conducts current when the voltage at OUT exceeds the voltage at IN. This current is not limited, so if extended reverse voltage operation is anticipated, external limiting to 5% of rated output current may be appropriate.

SHUTDOWN

The enable pin (EN) is active high and is compatible with standard and low voltage, TTL-CMOS levels. When shutdown capability is not required, EN can be connected to the IN pin, as shown in Figure 40. Figure 41 shows when both EN and VSET are tied to IN. The TPS728xx, with internal active output pulldown circuitry, discharges the output to within 5% of V_{OUT} with a time (*t*) of:

$$t = 3 \left[\frac{60 \times R_L}{60 + R_L} \right] \times C_{OUT}$$

Where:

 R_L = output load resistance C_{OUT} = output capacitance

DROPOUT VOLTAGE

The TPS728xx uses a PMOS pass transistor to achieve low dropout. When $(V_{IN}-V_{OUT})$ is less than the dropout voltage (V_{DO}) , the PMOS pass device is in the linear region of operation and the input-to-output resistance is the $R_{DS(ON)}$ of the PMOS pass element. V_{DO} approximately scales with output current because the PMOS device behaves like a resistor in dropout.

As with any linear regulator, PSRR and transient response are degraded as $(V_{\text{IN}}-V_{\text{OUT}})$ approaches dropout. This effect is shown in Figure 25 and Figure 26 in the Typical Characteristics section.

TRANSIENT RESPONSE

As with any regulator, increasing the size of the output capacitor reduces over/undershoot magnitude but increases duration of the transient response.

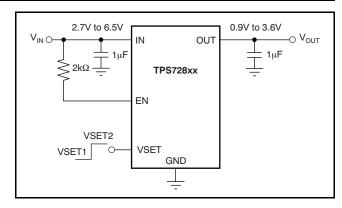


Figure 40. Circuit Showing EN Tied High when Shutdown Capability is Not Required

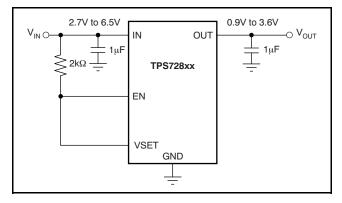


Figure 41. Circuit to Tie Both EN and VSET High

UNDERVOLTAGE LOCK-OUT (UVLO)

The TPS728xx uses an undervoltage lock-out circuit to keep the output shut off until the internal circuitry is operating properly. The UVLO circuit has a deglitch feature so that it typically ignores undershoot transients on the input if they are less than 5µs duration. The UVLO circuit triggers at approximately 2.3V on an undershooting or a falling input voltage. On the TPS728xx, the active pulldown discharges V_{OUT} when the device is in UVLO off condition. However, the input voltage must be greater than 0.8V for the active pulldown to work.

MINIMUM LOAD

The TPS728xx is stable with no output load. Traditional PMOS LDO regulators suffer from lower loop gain at very light output loads. The TPS728xx employs an innovative, low-current mode circuit under very light or no-load conditions, resulting in improved output voltage regulation performance down to zero output current.



THERMAL INFORMATION

Thermal Protection

Thermal protection disables the output when the junction temperature rises to approximately +160°C, allowing the device to cool. When the junction temperature cools to approximately +140°C the output circuitry is again enabled. Depending on power dissipation, thermal resistance, and ambient temperature, the thermal protection circuit may cycle on and off. This cycling limits the dissipation of the regulator, protecting it from damage as a result of overheating.

Any tendency to activate the thermal protection circuit indicates excessive power dissipation or an inadequate heatsink. For reliable operation, junction temperature should be limited to +125°C maximum. To estimate the margin of safety in a complete design heatsink), (including increase the temperature until the thermal protection is triggered; use worst-case loads and signal conditions. For good reliability, thermal protection should trigger at least +35°C above the maximum expected ambient condition of your particular application. configuration produces a worst-case junction temperature of +125°C at the highest expected ambient temperature and worst-case load.

The internal protection circuitry of the TPS728xx has been designed to protect against overload conditions.

It was not intended to replace proper heatsinking. Continuously running the TPS728xx into thermal shutdown degrades device reliability.

Power Dissipation

The ability to remove heat from the die is different for each package type, presenting different considerations in the printed circuit board (PCB) layout. The PCB area around the device that is free of other components moves the heat from the device to the ambient air. Performance data for JEDEC lowand high-K boards are given in the *Dissipation Ratings* table. Using heavier copper increases the effectiveness in removing heat from the device. The addition of plated through-holes to heat-dissipating layers also improves the heatsink effectiveness.

Power dissipation depends on input voltage and load conditions. Power dissipation (P_D) is equal to the product of the output current times the voltage drop across the output pass element (V_{IN} to V_{OUT}), as shown in Equation 1:

$$P_{D} = (V_{IN} - V_{OUT}) \times I_{OUT}$$
 (1)

Package Mounting

Solder pad footprint recommendations for the TPS728xx are available from the Texas Instruments web site at www.ti.com.

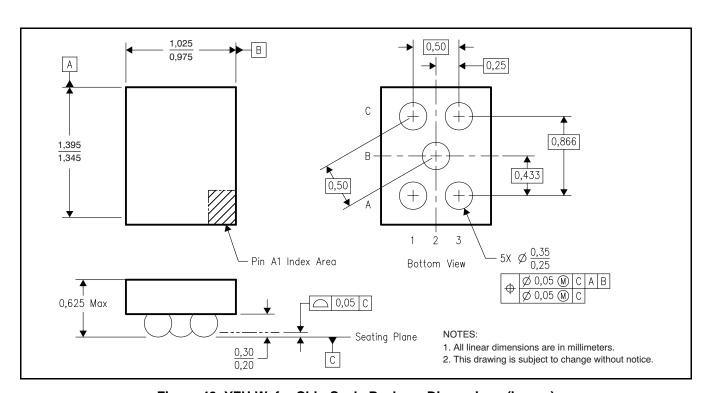


Figure 42. YZU Wafer Chip-Scale Package Dimensions (in mm)





com 3-Nov-2008

PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	e Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
TPS728185315DRVR	ACTIVE	SON	DRV	6	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS728185315DRVRG4	ACTIVE	SON	DRV	6	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS728185315DRVT	ACTIVE	SON	DRV	6	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS728185315DRVTG4	ACTIVE	SON	DRV	6	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS728185315YZUR	ACTIVE	DSBGA	YZU	5	3000	Green (RoHS & no Sb/Br)	Call TI	Level-1-260C-UNLIM
TPS728185315YZUT	ACTIVE	DSBGA	YZU	5	250	Green (RoHS & no Sb/Br)	Call TI	Level-1-260C-UNLIM

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

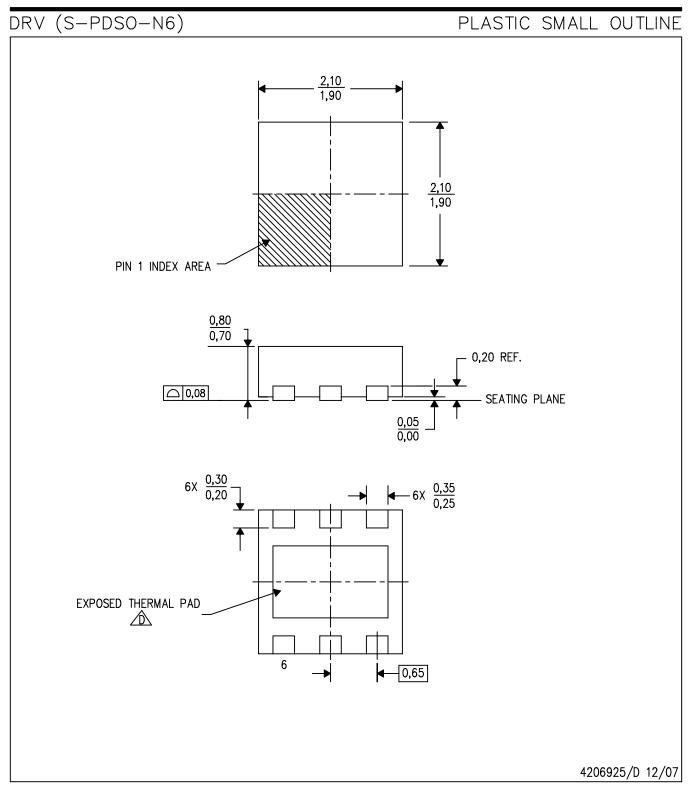
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS728185315DRVR	SON	DRV	6	3000	179.0	8.4	2.2	2.2	1.2	4.0	8.0	Q2
TPS728185315DRVT	SON	DRV	6	250	179.0	8.4	2.2	2.2	1.2	4.0	8.0	Q2
TPS728185315YZUR	DSBGA	YZU	5	3000	178.0	8.4	1.09	1.42	0.81	4.0	8.0	Q1
TPS728185315YZUT	DSBGA	YZU	5	250	178.0	8.4	1.09	1.42	0.81	4.0	8.0	Q1





*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS728185315DRVR	SON	DRV	6	3000	195.0	200.0	45.0
TPS728185315DRVT	SON	DRV	6	250	195.0	200.0	45.0
TPS728185315YZUR	DSBGA	YZU	5	3000	217.0	193.0	35.0
TPS728185315YZUT	DSBGA	YZU	5	250	217.0	193.0	35.0



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

- B. This drawing is subject to change without notice.
- C. Small Outline No-Lead (SON) package configuration.

The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.



THERMAL PAD MECHANICAL DATA



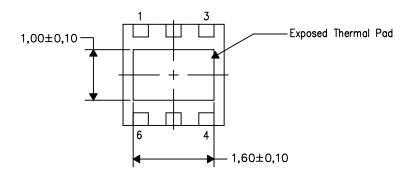
DRV (S-PDSO-N6)

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No—Lead (QFN) package and its advantages, refer to Application Report, Quad Flatpack No—Lead Logic Packages, Texas Instruments Literature No. SCBA017. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.

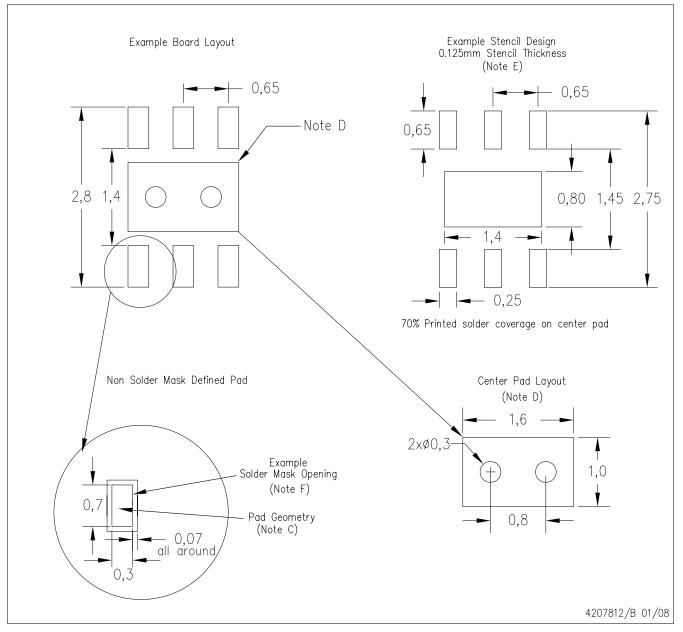


Bottom View

NOTE: All linear dimensions are in millimeters

Exposed Thermal Pad Dimensions

DRV (S-PDSO-N6)



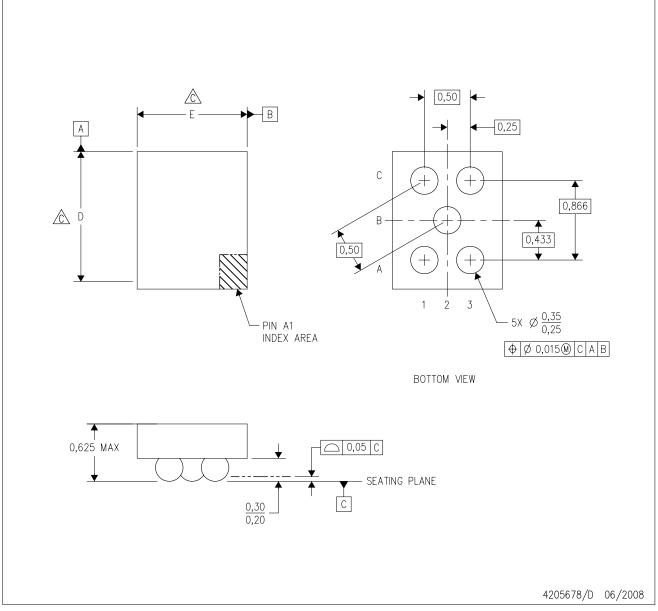
NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, QFN Packages, Texas Instruments Literature No. SCBA017, SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com www.ti.com.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- F. Customers should contact their board fabrication site for solder mask tolerances.



YZU (R-XBGA-N5)

DIE-SIZE BALL GRID ARRAY



Notes:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- Devices in this YZQ package can have dimension D ranging from 1.31 to 1.75 mm and dimension E ranging from 0.94 to 1.45 mm.

 To determine the exact package size of a particular device, refer to the device datasheet or contact a local TI representative.
- D. NanoFree™ package configuration.
- E. This package contains lead-free balls. Refer to the 5 YEU package (drawing 4205430) for tin-lead (SnPb) balls.

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